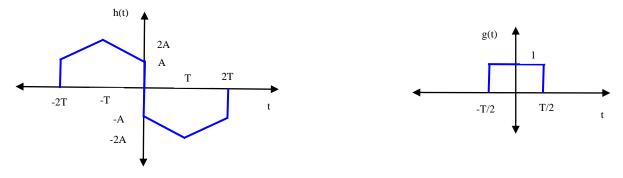
## Department of Electrical Engineering Indian Institute of Technology, Madras

## **EE 3005: Communication Systems**

February 22, 2025 Tutorial #2 KG/IITM
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**1.** Express the Fourier transform H(f) of h(t) in terms of the Fourier transform G(f) when g(t) and h(t) are given as below:



2. A real, bandpass, QCM signal  $s(t) = m_1(t)Cos(2\pi f_c t) + m_2(t)Sin(2\pi f_c t)$  with carrier frequency  $f_c$ =200KHz conveys two message signals  $m_1(t)$  and  $m_2(t)$  of <u>one-sided</u> bandwidth of W=5KHz each. Let S(f),  $M_1(f)$ , and  $M_2(f)$  denote the (complex) Fourier transforms of s(t),  $m_1(t)$  and  $m_2(t)$ , respectively. Given  $M_1(f=2KHz) = a + jb$  and  $M_2(f=2KHz) = c + jd$  where j = sqrt(-1), determine the following in terms of *a*, *b*, *c*, and *d*. The number inside the brackets can be assumed to be in KHz. (a)  $M_1(-2)$  and  $M_2(-2)$ (b) S(202) and S(198)(c) S(-198)

**3.** Prove that the angle-modulated signal  $s(t) = e^{-\hat{\varphi}(t)} Cos(2\pi f_C t + \varphi(t))$  has no frequency components in  $-f_C < f < f_C$ , where  $\hat{\varphi}(t)$  is the Hilbert transform of  $\varphi(t)$ .

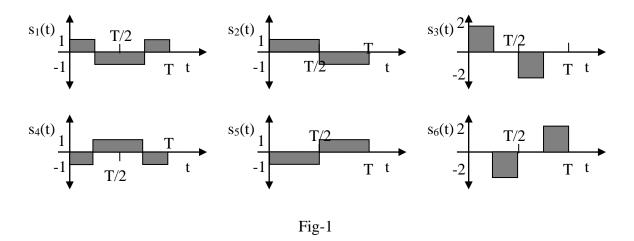
*Hints for the proof*: First show that this signal can be rewritten as  $Re\{e^{j(2\pi f_C t + \varphi_+(t))}\}$  where  $\varphi_+(t)$  is the pre-envelope signal. Then use the fact that for a complex quantity *z*, one can write  $Re\{z\} = \frac{1}{2}(z + z^*)$  to get an alternate expression for s(t). Then, use the series expansion  $e^{j\varphi_+(t)} = \sum_{n=0}^{\infty} \frac{j^n}{n!} \varphi_+^n(t)$  and take Fourier transform on both sides. You should finally get the result by recalling the property about the FT of the pre-envelope signal.

4. The following questions from Chapter 3 of the text book (E-version), pp-143 onwards:

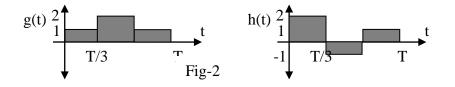
**3.14** to **3.16**, and **3.17**\*. *\* these questions could have a higher order of difficulty to solve* 

5. <u>Reading exercise</u>: Up to Section 4.1 in the textbook (up to page 159). Attempt question 4.18.

**6.** Find the compact ortho-normal basis set, and using it, make a clear labeled plot of the signal constellation for the signal set shown in Fig-1.



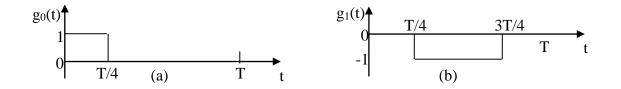
7. The signal g(t) is sent through a channel with impulse response h(t), where the two functions are shown in Fig-2. Make a labeled plot of the ideal matched filter's impulse response. *Hint:* Assume single-shot communication.



**8.** Find the autocorrelation function S(t) for the given signal g(t) in Fig3. Draw the signal S(t).

where  $S(t) = \int_{-\infty}^{\infty} g(\tau)g(t-\tau)d\tau$  g(t) fig-3 g(t) g(t) fig-3

9. Find out the compact basis function for signals given in Fig-4.



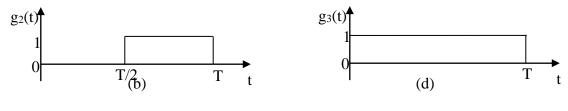


Fig-4

10. Consider the signal set shown in Fig-5 below.

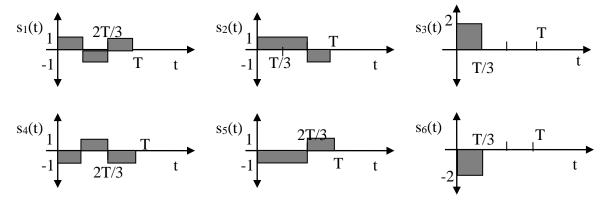


Fig-5

(a) Find a compact orthonormal basis set for this signal set. Sketch these functions.

(b) Using this, make a clear labeled plot of the corresponding signal constellation.

(c) In terms of the average energy  $E_a$  of the constellation, what is the minimum distance (i.e., 2d) of the signal set?

**11.** Find the minimum distance and multiplicity for given signals in Fig-6.

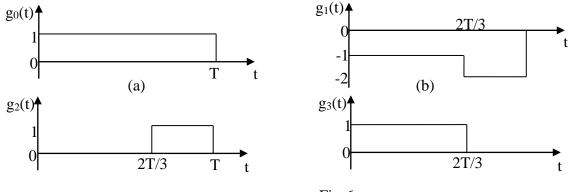


Fig-6

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